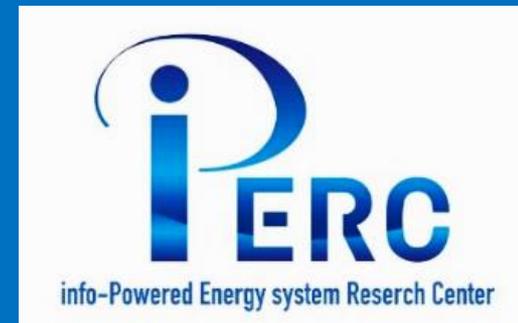




i-PERC
電気通信大学



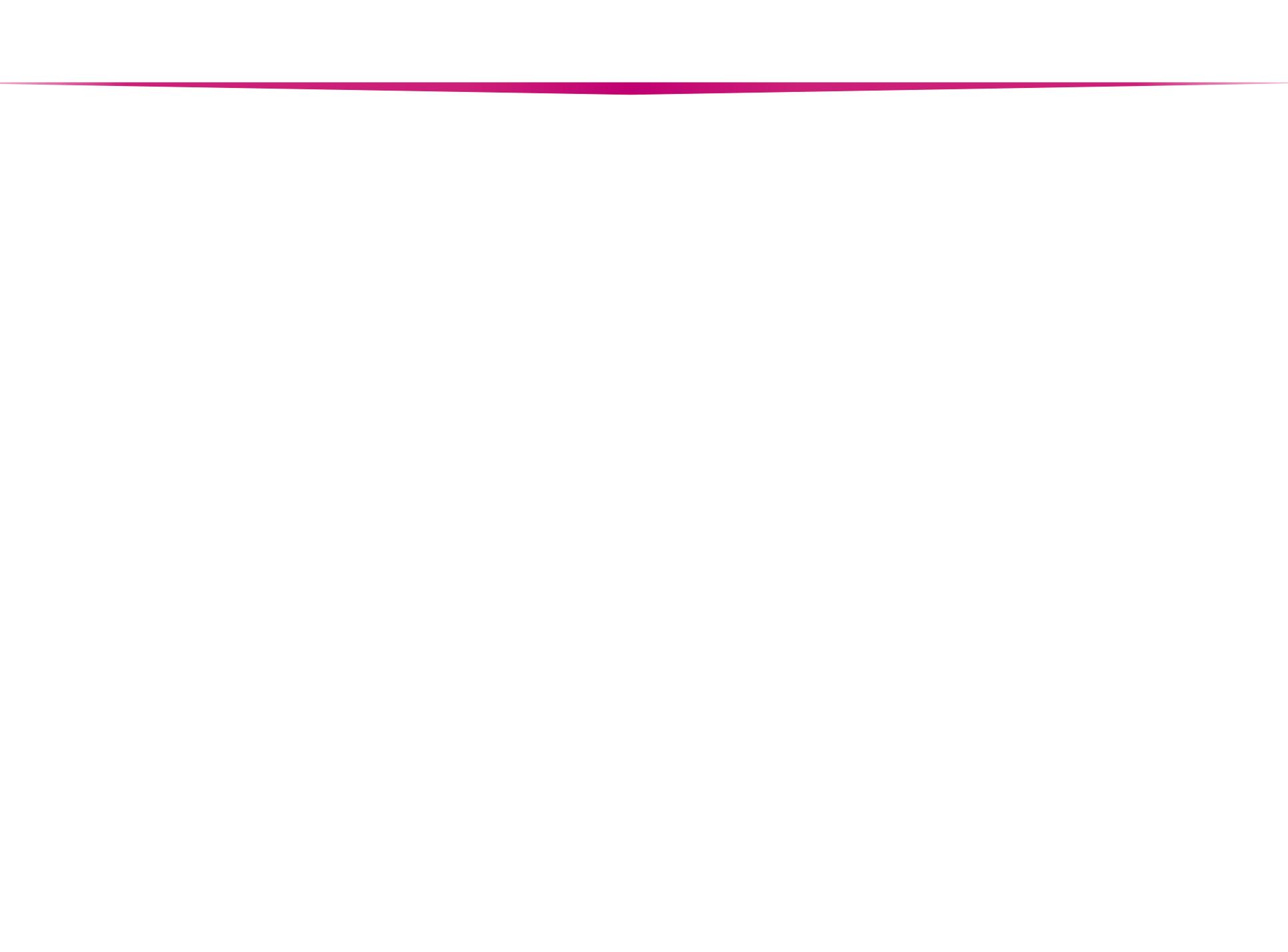
基礎電気回路CH-11

相互誘導回路

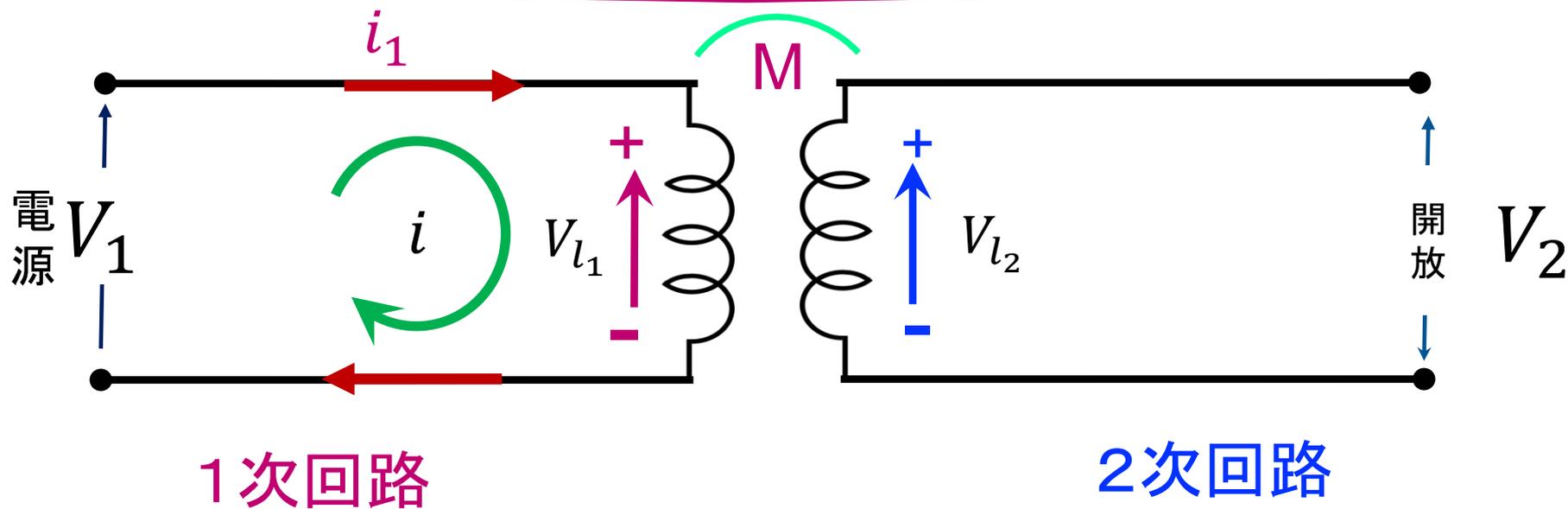
等価回路

変圧器回路

理想変圧器



相互誘導インダクタンス回路にKVL法則を応用する



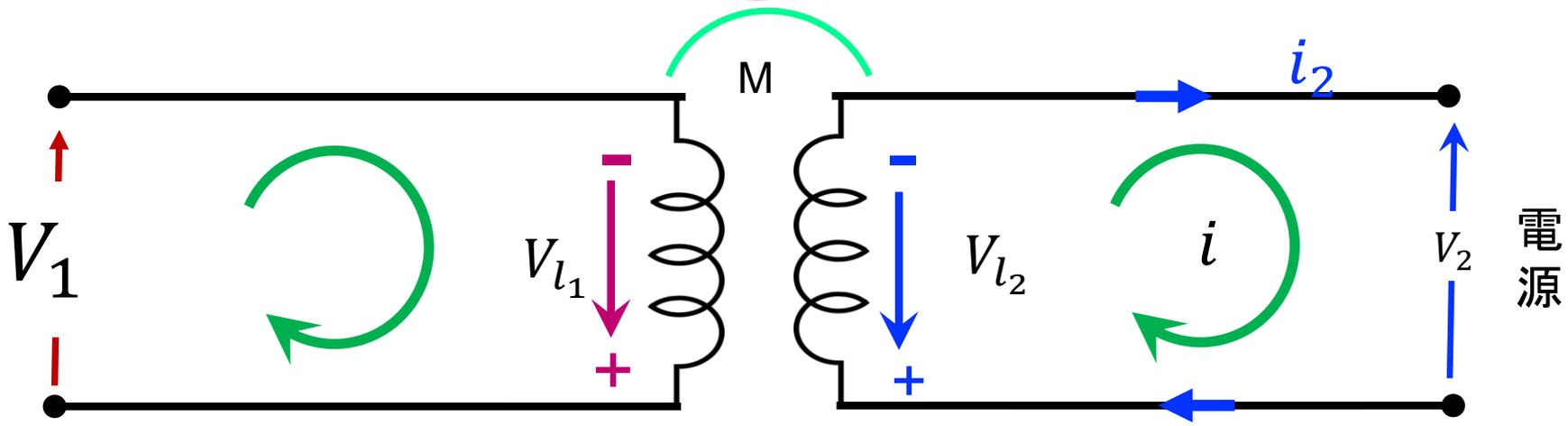
$$V_{l_1} - V_1 = 0$$

$$V_2 - V_{l_2} = 0$$

$$V_1 = V_{l_1} = L_1 \frac{di_1}{dt}$$

$$V_2 = V_{l_2} = M_{21} \frac{di_1}{dt}$$

相互誘導インダクタンス回路にKVL法則を応用する



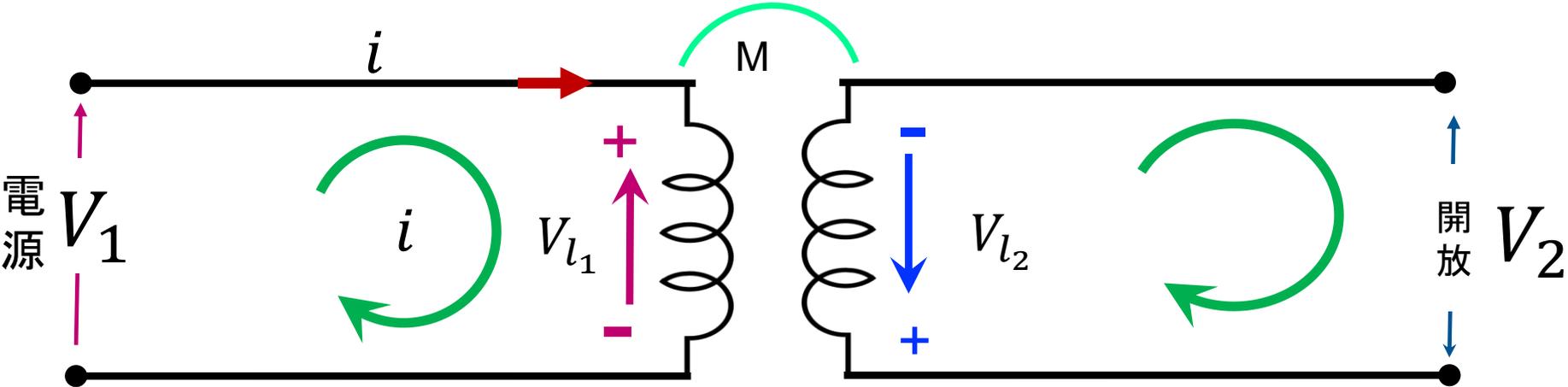
2次回路

1次回路

$$-V_{l_1} + V_1 = 0$$
$$V_1 = V_{l_1} = M_{12} \frac{di_2}{dt}$$

$$-V_2 + V_{l_2} = 0$$
$$V_2 = V_{l_2} = L_2 \frac{di_2}{dt}$$

相互誘導インダクタンス回路にKVL法則を応用する: 逆巻き



1次回路

2次回路

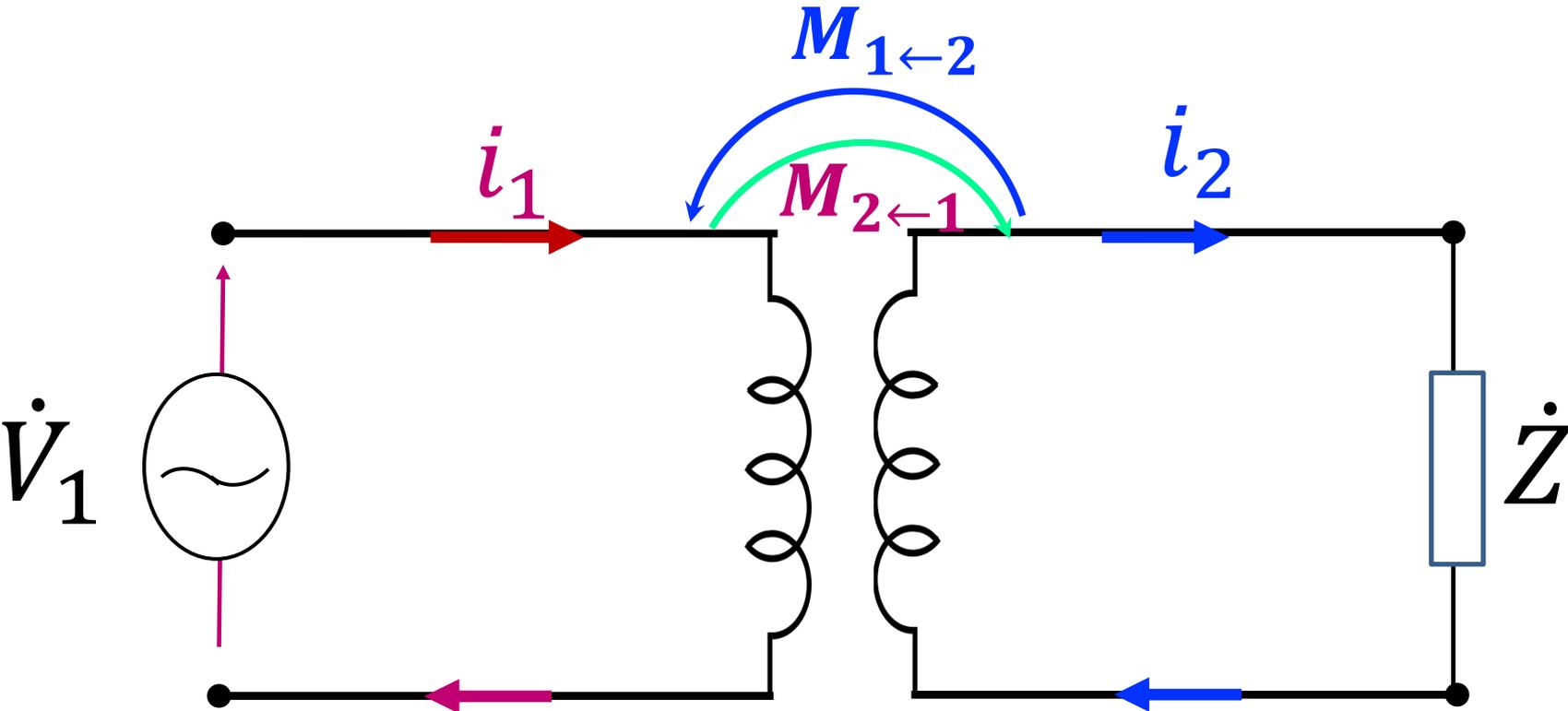
$$V_{l_1} - V_1 = 0$$

$$V_2 + V_{l_2} = 0$$

$$V_1 = V_{l_1} = L_1 \frac{di_1}{dt}$$

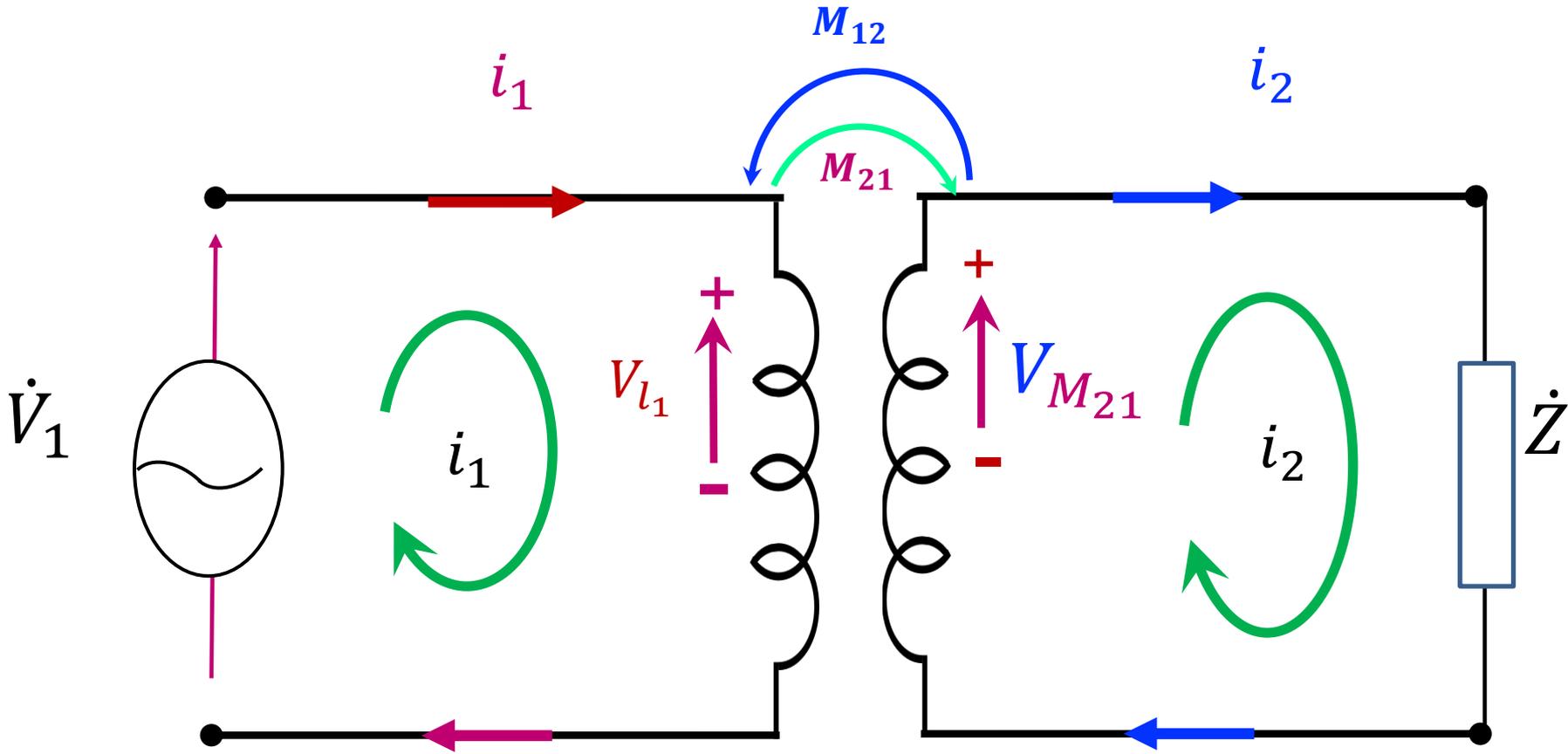
$$V_2 = -V_{l_2} = M_{21} \frac{di_1}{dt}$$

電磁誘導結合回路の拡張：交流電源と負荷を入れる



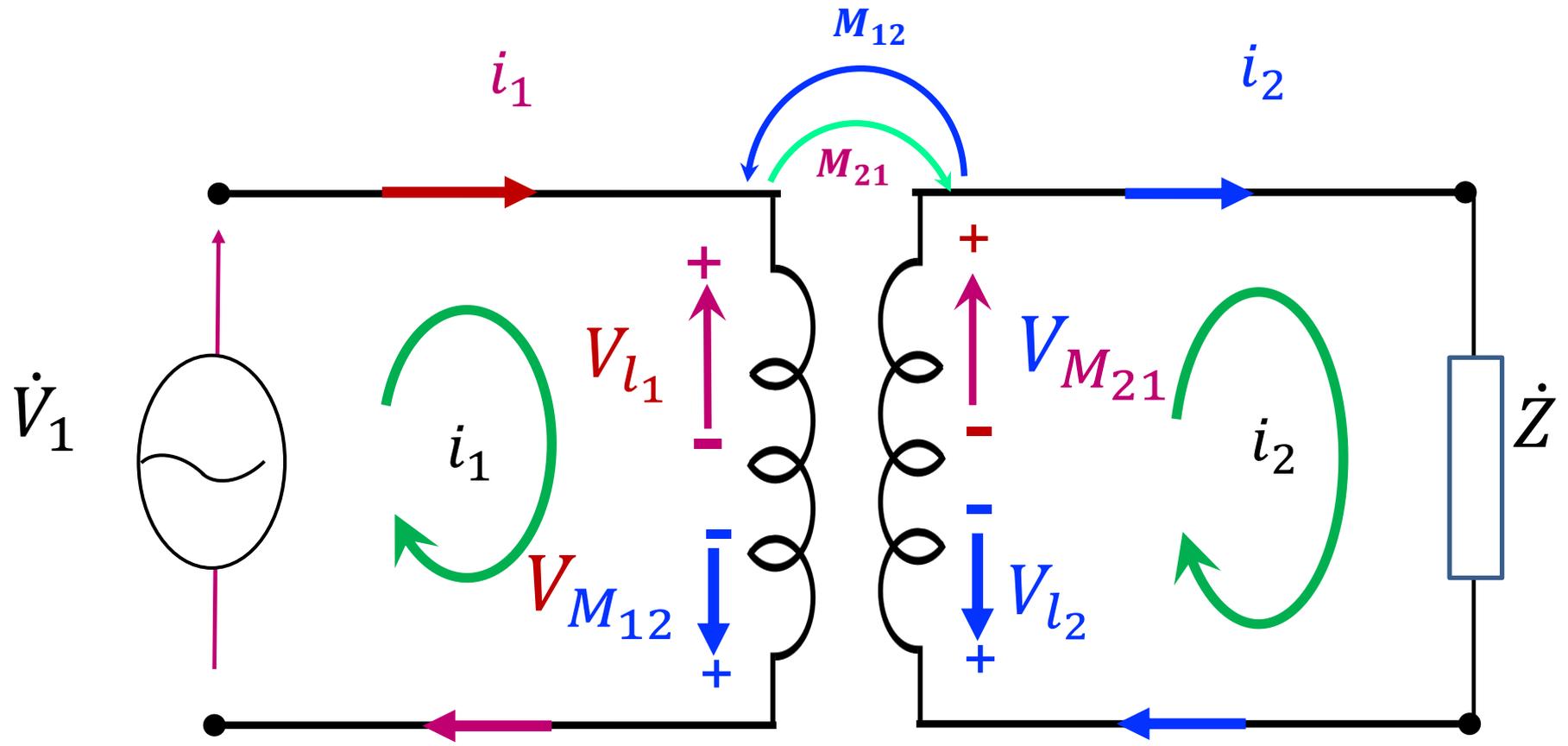
電磁誘導結合回路：KVL法則

$$M = M_{12} = M_{21}$$

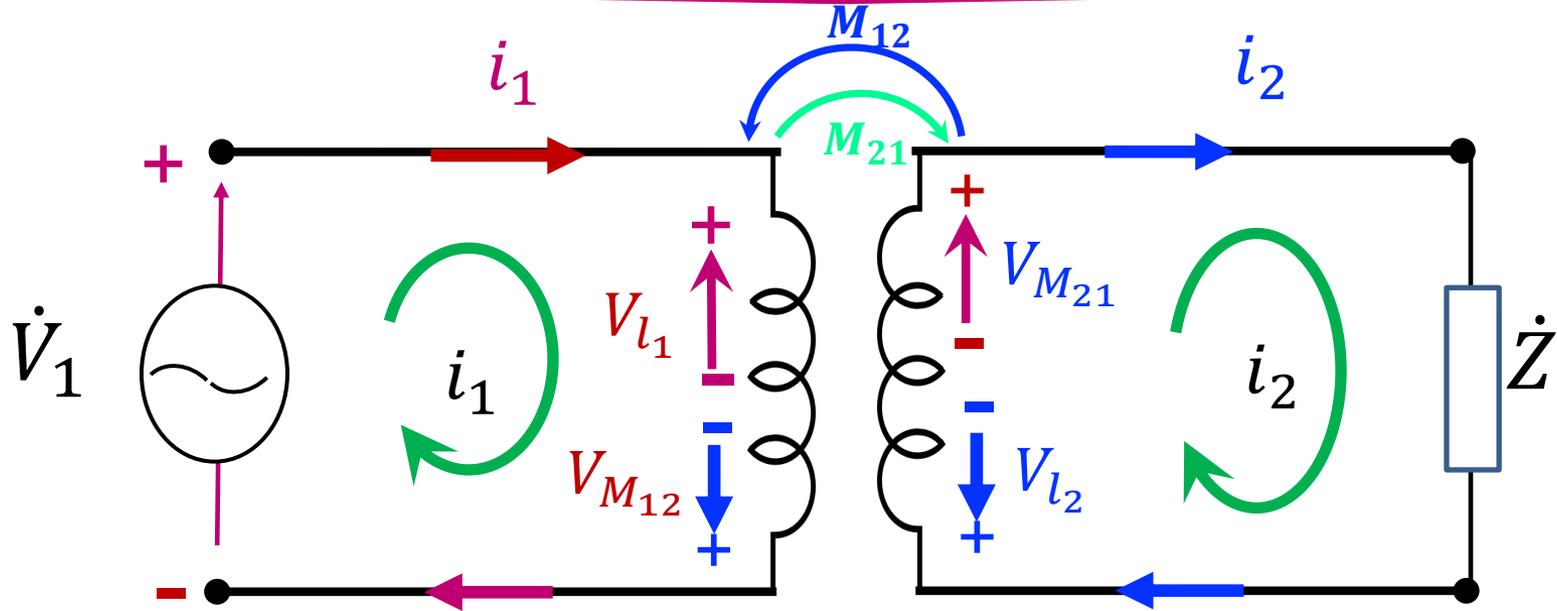


電磁誘導結合回路：KVL法則

$$M = M_{12} = M_{21}$$



電磁誘導結合回路：ループ1におけるKVL法則



(a) $V_{l_1} + V_{M_{12}} - \dot{V}_1 = 0$

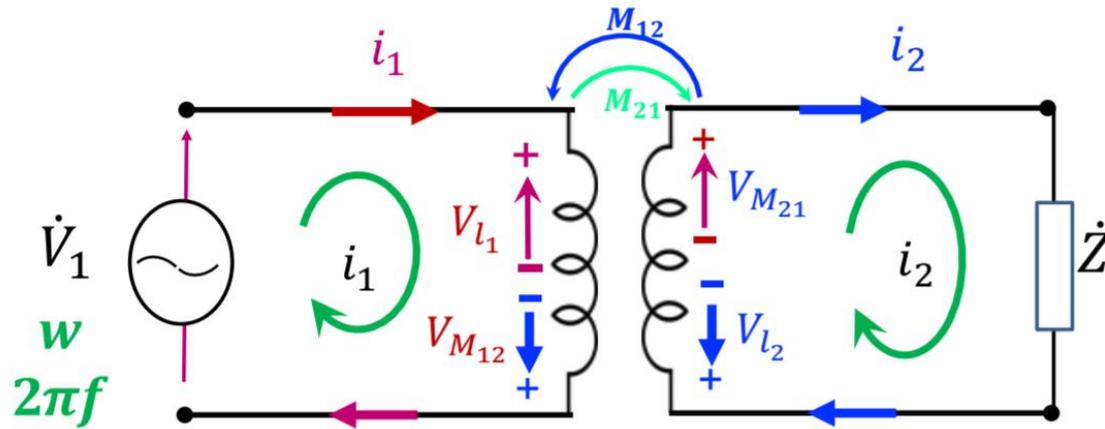
(c) $V_{l_1} - V_{M_{12}} - \dot{V}_1 = 0$

(b) $V_{l_1} + V_{M_{12}} + \dot{V}_1 = 0$

(d) $V_{l_1} - V_{M_{12}} + \dot{V}_1 = 0$

正解は:

電磁誘導結合回路：KVL法則

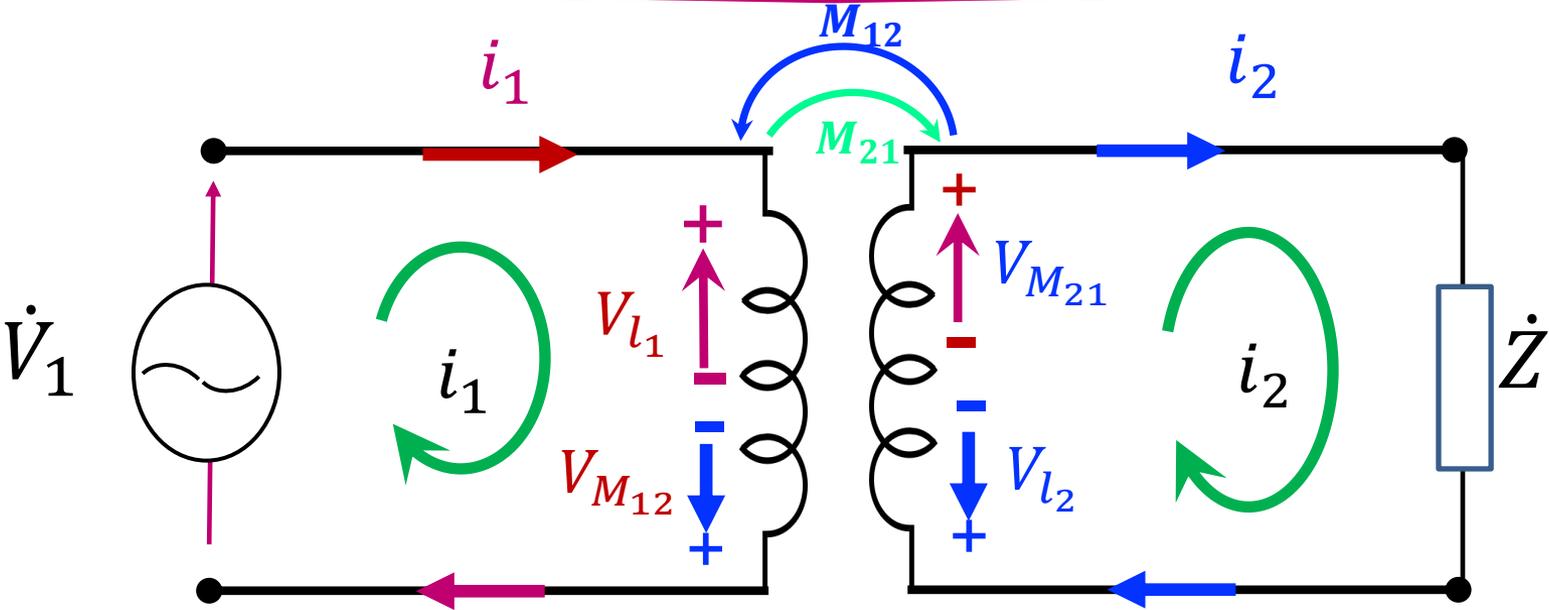


$$V_{L1} - V_{M12} - \dot{V}_1 = 0$$

$$L_1 \frac{di_1}{dt} - M_{12} \frac{di_2}{dt} - \dot{V}_1 = 0$$

$$j\omega L_1 i_1 - j\omega M_{12} i_2 = \dot{V}_1$$

電磁誘導結合回路：KVL法則



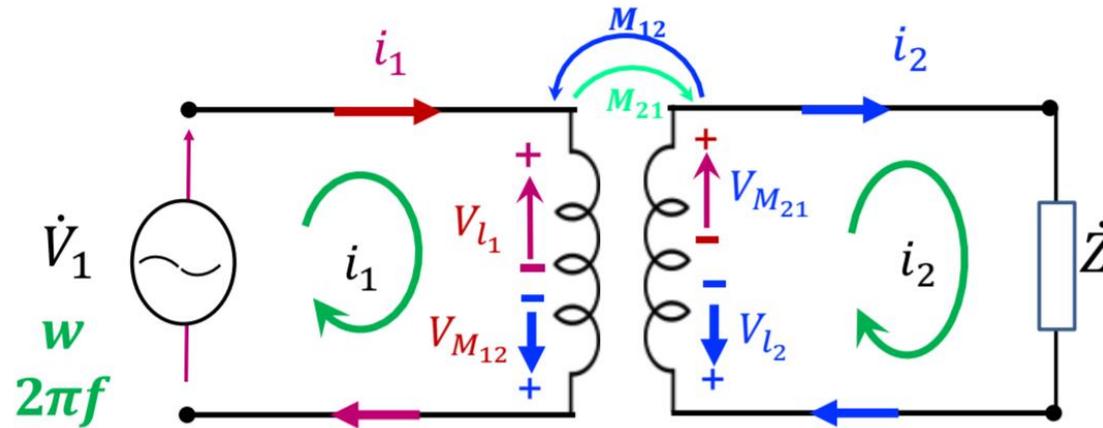
ループ2におけるKVL法則:

(a) $i_2 \dot{Z} + V_{M21} - V_{L2} = 0$ (c) $V_{M21} - V_{L2} + i_2 \dot{Z} = 0$

(b) $-V_{M21} + V_{L2} + i_2 \dot{Z} = 0$ (d) $V_{L2} + V_{M21} - i_2 \dot{Z} = 0$

正解は:

電磁誘導結合回路：KVL法則

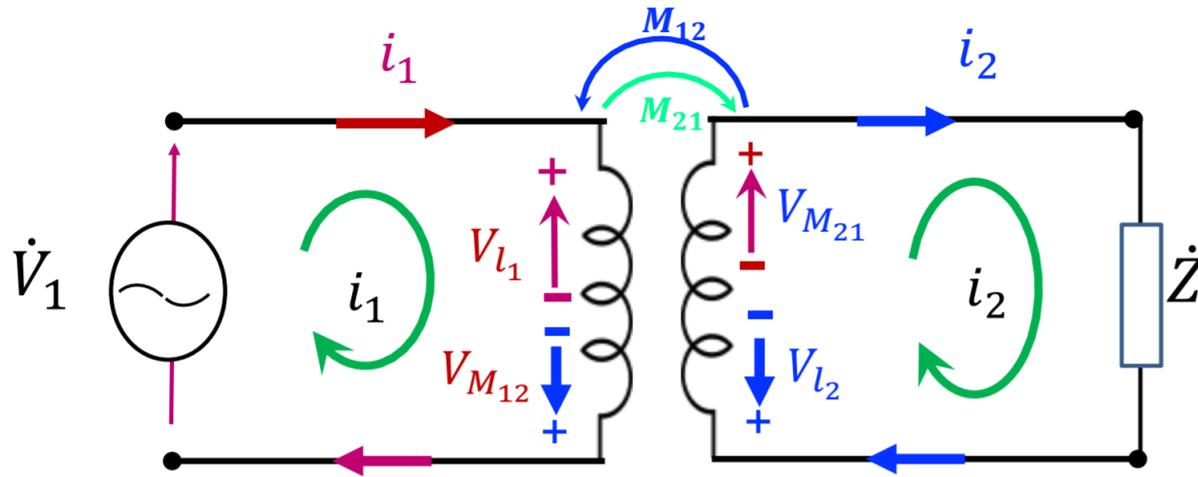


$$-V_{M_{21}} + V_{L_2} + i_2 \dot{Z} = 0$$

$$M_{21} \frac{di_1}{dt} - L_2 \frac{di_2}{dt} = \dot{Z} i_2$$

$$j\omega M_{12} i_1 - j\omega L_2 i_2 = \dot{Z} i_2$$

電磁誘導結合回路：KVL法則



1次回路

$$\dot{V}_1 = j\omega L_1 i_1 - j\omega M_{12} i_2$$

2次回路

$$\dot{Z} i_2 = j\omega M_{21} i_1 - j\omega L_2 i_2$$

電磁誘導結合回路：1次回路におけるインピーダンス

$$M = M_{12} = M_{21}$$

$$\dot{Z}i_2 = j\omega M_{21} i_1 - j\omega L_2 i_2$$

$$i_2 = \frac{j\omega M}{j\omega L_2 + \dot{Z}} i_1$$

$$\dot{V}_1 = j\omega L_1 i_1 - j\omega M_{12} i_2$$

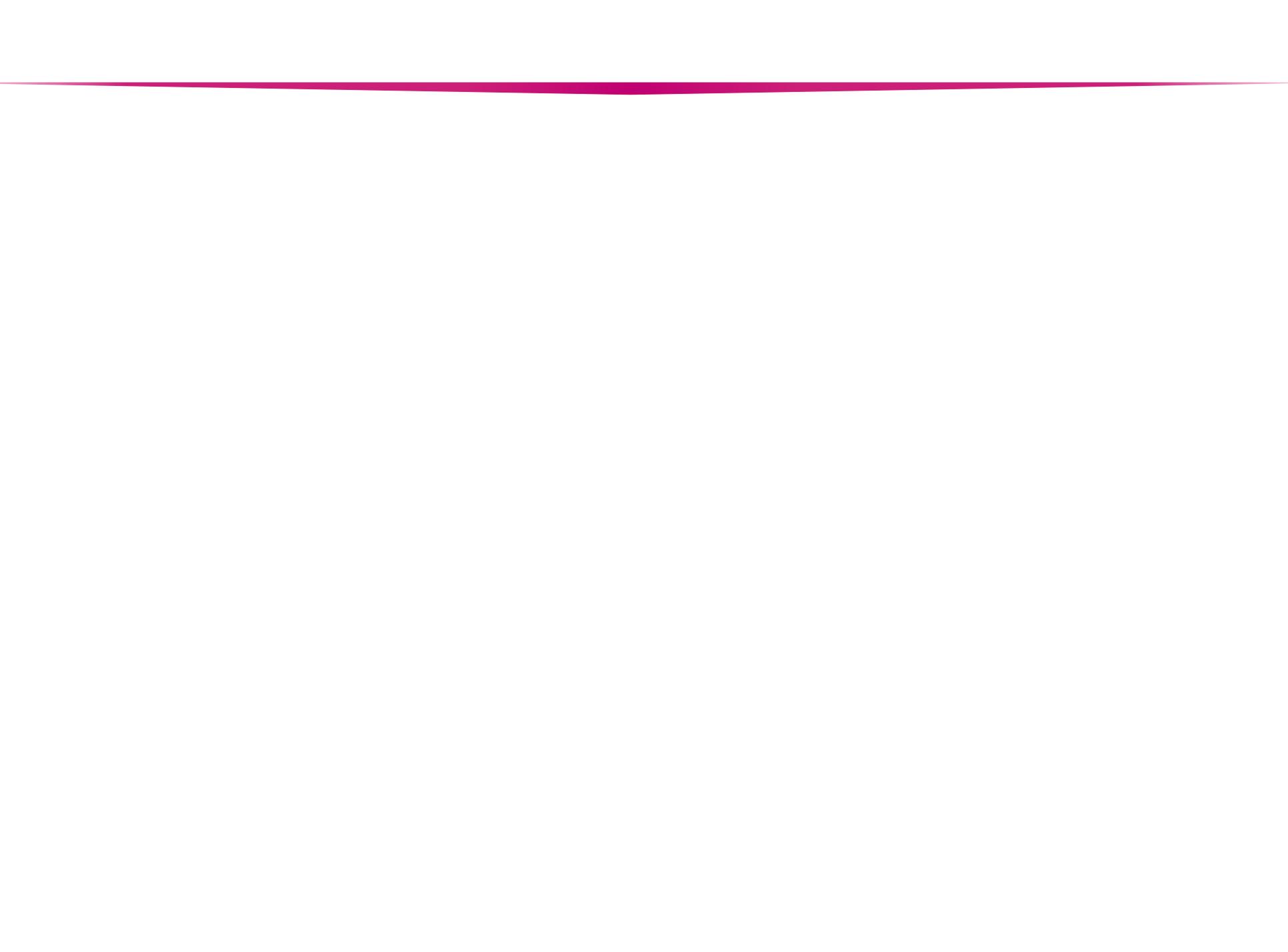
$$\dot{V}_1 = j\omega L_1 i_1 - j\omega M \frac{j\omega M}{j\omega L_2 + \dot{Z}} i_1$$

電磁誘導結合回路：1次回路におけるインピーダンス

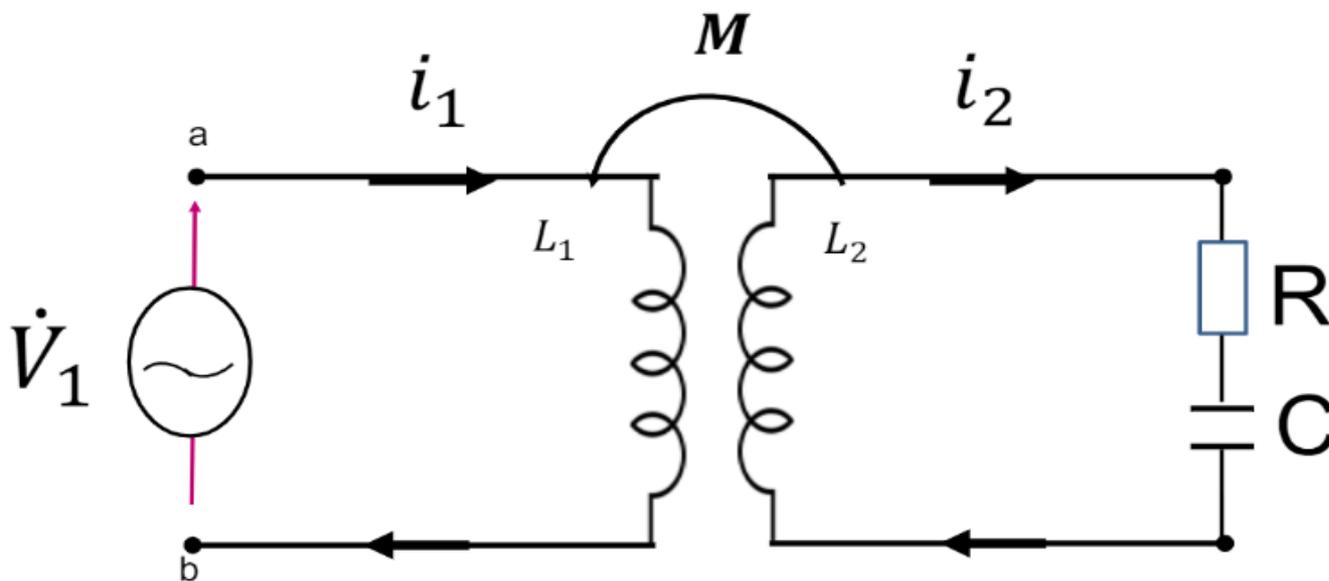
$$\dot{V}_1 = j\omega L_1 \dot{i}_1 - j\omega M \frac{j\omega M}{j\omega L_2 + \dot{Z}} \dot{i}_1$$

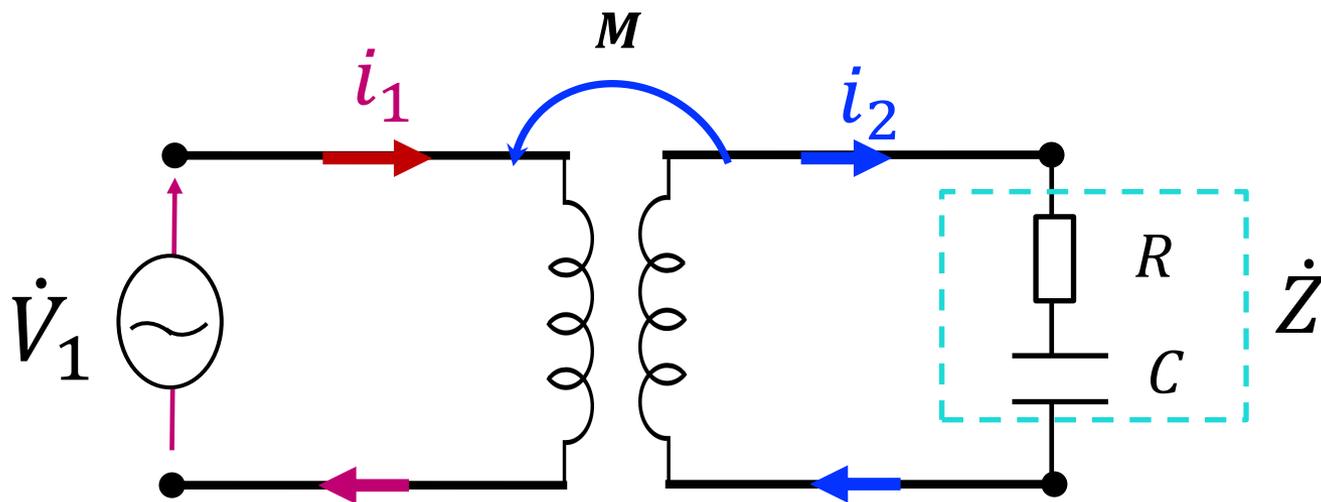
$$\dot{V}_1 = \left\{ j\omega L_1 + \frac{\omega^2 M^2}{j\omega L_2 + \dot{Z}} \right\} \dot{i}_1$$

$$\dot{i}_1 = \frac{\dot{V}_1}{j\omega L_1 + \frac{\omega^2 M^2}{j\omega L_2 + \dot{Z}}}$$



下図の電磁誘導結合回路において、 $L_1=10[\text{mH}]$, $L_2=4[\text{mH}]$, $M=2[\text{mH}]$, $R=100[\Omega]$, $C=3[\mu\text{F}]$ 、周波数 $f=500[\text{Hz}]$ との時、1次回路側（端子a,b）から見たインピーダンス \dot{Z}_1 の複素数表示と極表示を求めなさい。また、1次回路側に $\dot{V}_1 = 20\angle 0^\circ[\text{V}]$ ($f=500[\text{Hz}]$)の電圧を加えた時の電流 i_1 , i_2 , 電圧 V_2 のフェーザ表示を求めなさい。





1次回路

$$\dot{V}_1 = j\omega L_1 i_1 - j\omega M_{12} i_2$$

2次回路

$$\dot{Z} i_2 = j\omega M_{21} i_1 - j\omega L_2 i_2$$

2次回路

$$\dot{Z}i_2 = j\omega M i_1 - j\omega L_2 i_2$$

$$\dot{Z} = R + \frac{1}{j\omega C}$$

$$i_2 = \frac{j\omega M}{R + \frac{1}{j\omega C} + j\omega L_2} i_1$$

$$i_2 = \frac{j\omega M}{R + j(\omega L_2 - \frac{1}{\omega C})} i_1$$

1次回路

$$\dot{V}_1 = j\omega L_1 i_1 - j\omega M i_2$$

$$\dot{V}_1 = j\omega L_1 i_1 - j\omega M \frac{j\omega M}{R + j(\omega L_2 - \frac{1}{\omega C})} i_1$$

$$\dot{V}_1 = j\omega L_1 i_1 + \frac{\omega^2 M^2}{R + j(\omega L_2 - \frac{1}{\omega C})} i_1$$

$$\dot{V}_1 = \left(j\omega L_1 + \frac{\omega^2 M^2}{R + j(\omega L_2 - \frac{1}{\omega C})} \right) i_1$$

$$\dot{Z}_1 = \frac{\dot{V}_1}{i_1} = j\omega L_1 + \frac{\omega^2 M^2}{R + j(\omega L_2 - \frac{1}{\omega C})}$$

これは一次側から見たインピーダンス \dot{Z}_1 である。

$$\omega = 2\pi f = 3140 \left(\frac{\text{rad}}{\text{s}} \right)$$

$$L_1 = 1 * 10^{-2} H \quad M = 2 * 10^{-3} H$$

$$L_2 = 4 * 10^{-3} H \quad R = 100 \Omega$$

$$C = 3 * 10^{-6} F$$

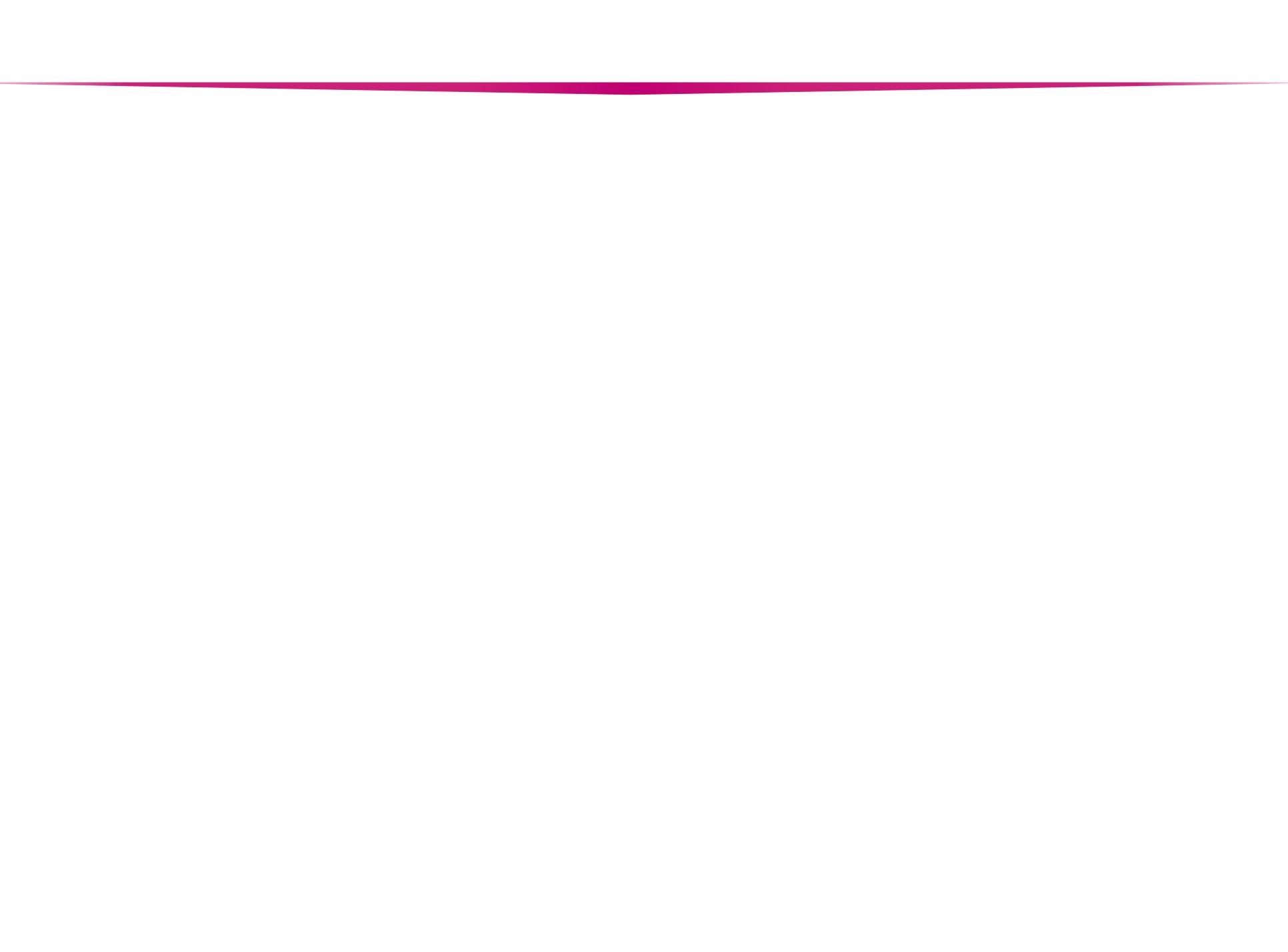
$$\dot{Z}_1 = j * 31.40 + \frac{3140^2 * 4 * 10^{-6}}{100 + j(3.14 * 4 - \frac{1}{3140 * 3 * 10^{-6}})}$$

$$\dot{Z}_1 = j * 31.40 + \frac{39.48}{100 - j(93.54)}$$

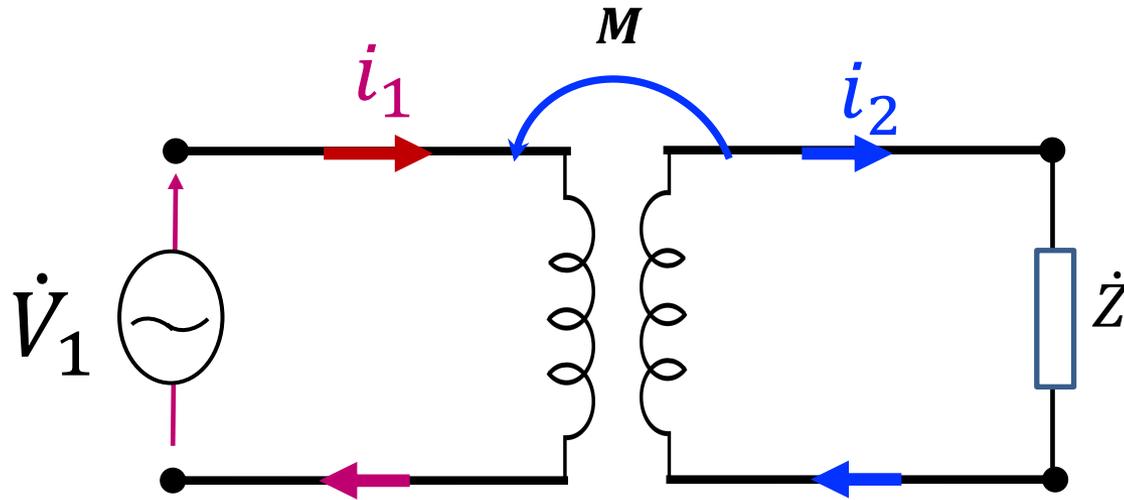
$$\dot{Z}_1 = j * 31.40 + \frac{39.48 + j3693}{18750} = 0.211 + j31.7$$

$$\dot{Z}_1 = 0.211 + j31.7 = \sqrt{0.211^2 + 31.7^2} \angle \tan^{-1}\left(\frac{31.7}{0.211}\right)$$

$$\dot{Z}_1 = 31.7 \angle 89.6^\circ$$



等価回路概念の導入



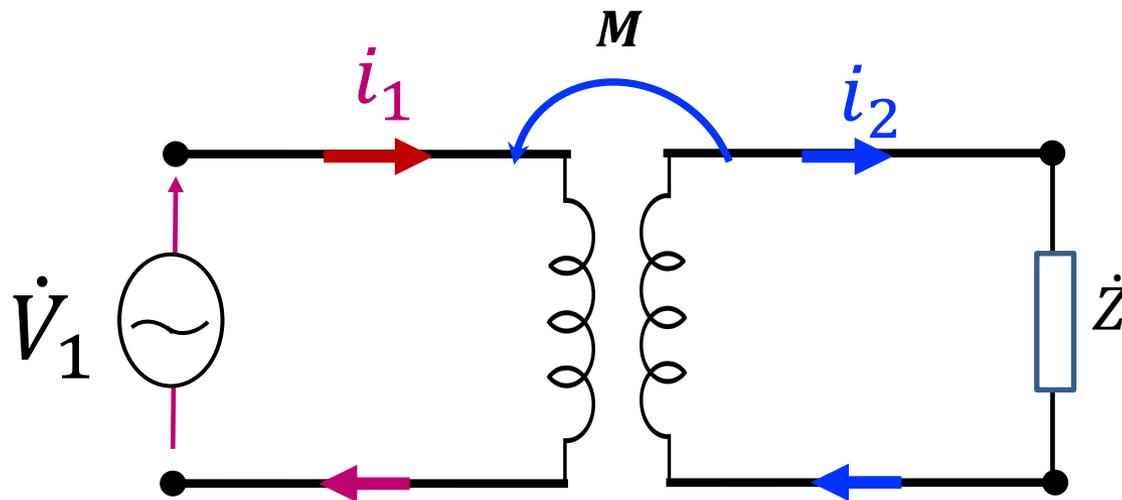
1次回路

$$\dot{V}_1 = j\omega L_1 i_1 - j\omega M i_2$$

2次回路

$$\dot{Z} i_2 = j\omega M i_1 - j\omega L_2 i_2$$

等価回路概念の導入



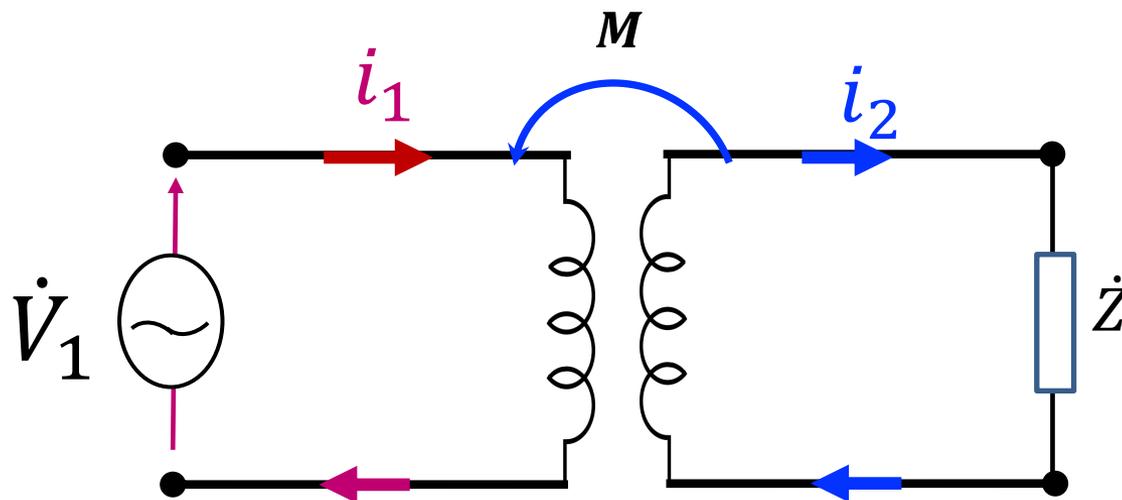
1次回路

$$\dot{V}_1 = j\omega L_1 i_1 - j\omega M i_2$$

$$\dot{V}_1 = j\omega(L_1 - M)i_1 + j\omega M i_1 - j\omega M i_2$$

$$\dot{V}_1 = j\omega(L_1 - M)i_1 + j\omega M(i_1 - i_2)$$

等価回路概念の導入



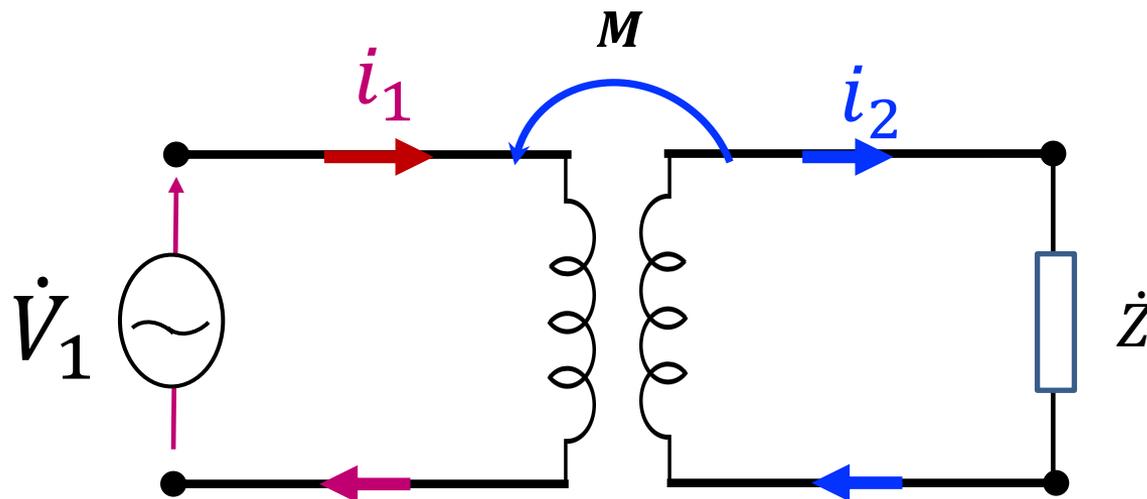
2次回路

$$\dot{Z}i_2 = j\omega M i_1 - j\omega L_2 i_2$$

$$\dot{Z}i_2 = j\omega M i_1 - j\omega M i_2 - j\omega(L_2 - M) i_2$$

$$\dot{Z}i_2 = j\omega M (i_1 - i_2) - j\omega(L_2 - M) i_2$$

等価回路概念の導入



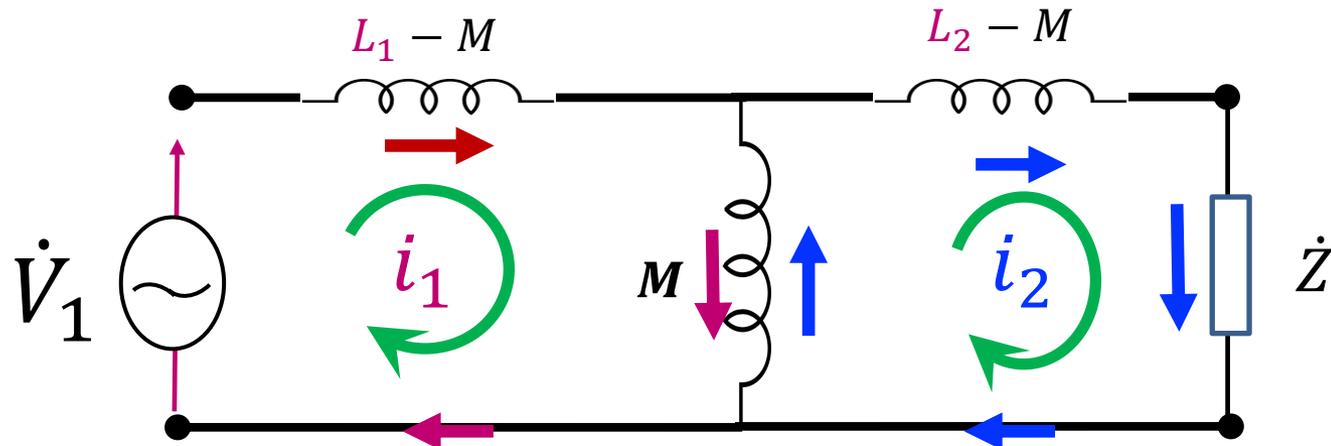
1次回路

$$\dot{V}_1 = j\omega(L_1 - M)i_1 + j\omega M(i_1 - i_2)$$

2次回路

$$\dot{Z}i_2 = j\omega M(i_1 - i_2) - j\omega(L_2 - M)i_2$$

等価回路概念の導入



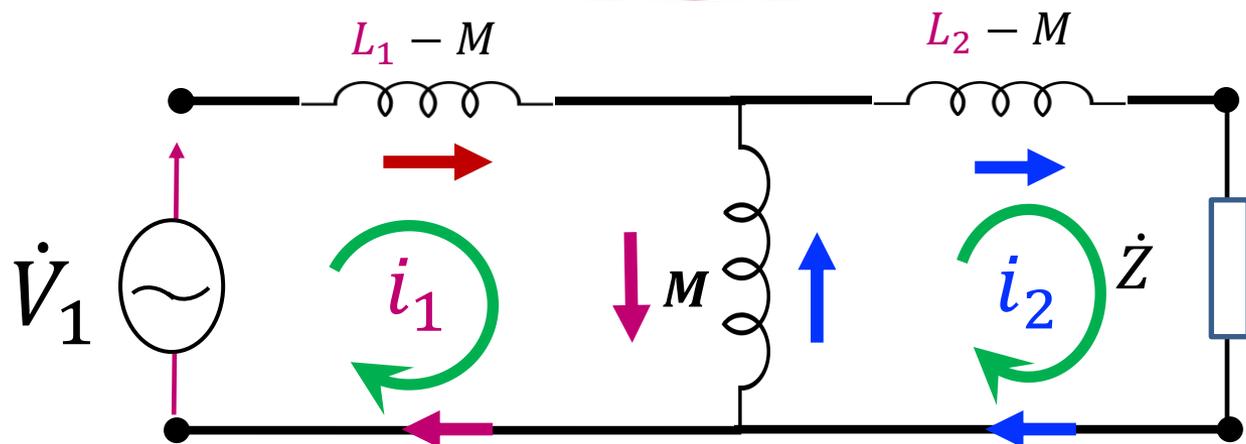
1次回路

$$\dot{V}_1 = j\omega(L_1 - M)i_1 + j\omega M(i_1 - i_2)$$

2次回路

$$\dot{Z}i_2 = j\omega M(i_1 - i_2) - j\omega(L_2 - M)i_2$$

等価回路概念の導入



1次回路

$$\dot{V}_1 = j\omega(L_1 - M)i_1 + j\omega M(i_1 - i_2)$$

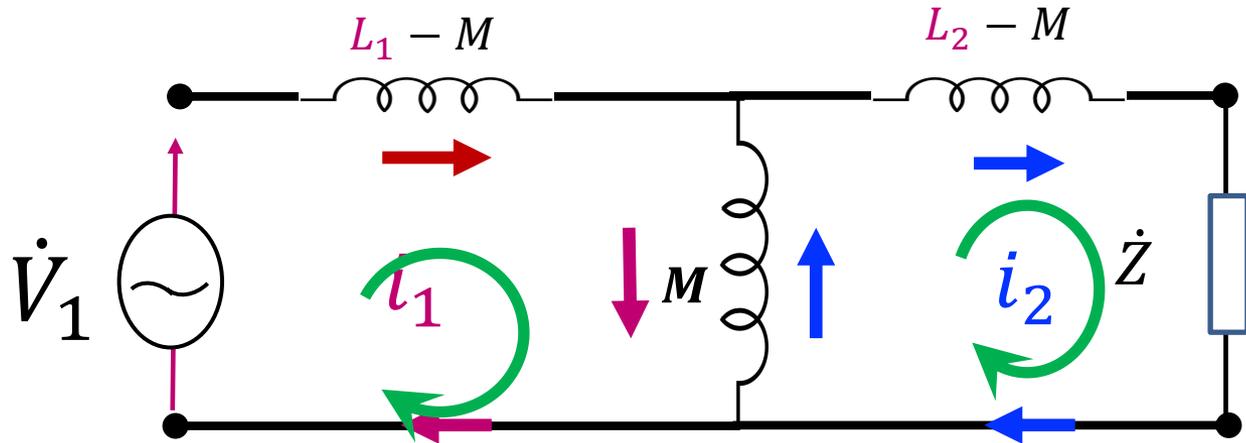
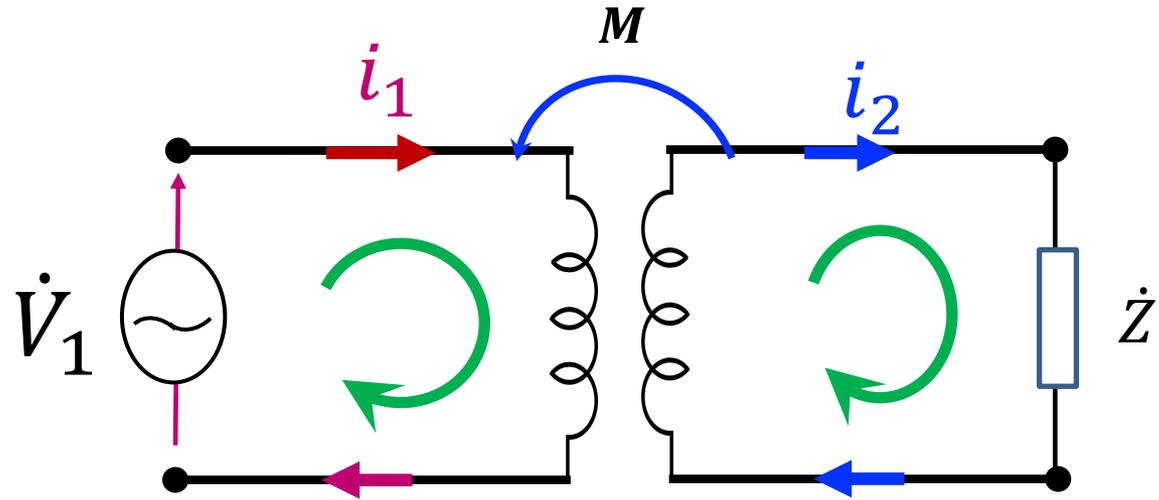
$$-\dot{V}_1 + j\omega(L_1 - M)i_1 + j\omega M(i_1 - i_2) = 0$$

2次回路

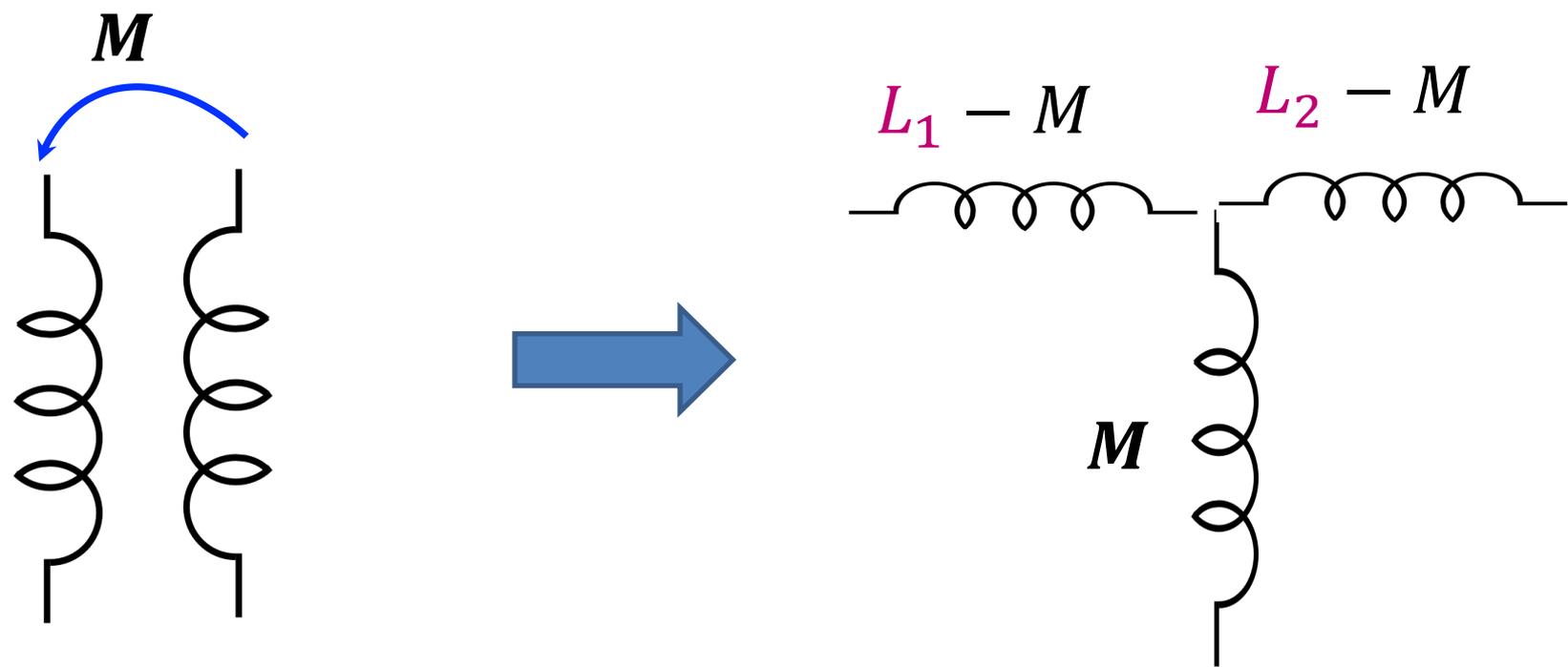
$$-\dot{Z}i_2 = j\omega M(i_2 - i_1) + j\omega(L_2 - M)i_2$$

$$\dot{Z}i_2 + j\omega M(i_2 - i_1) + j\omega(L_2 - M)i_2 = 0$$

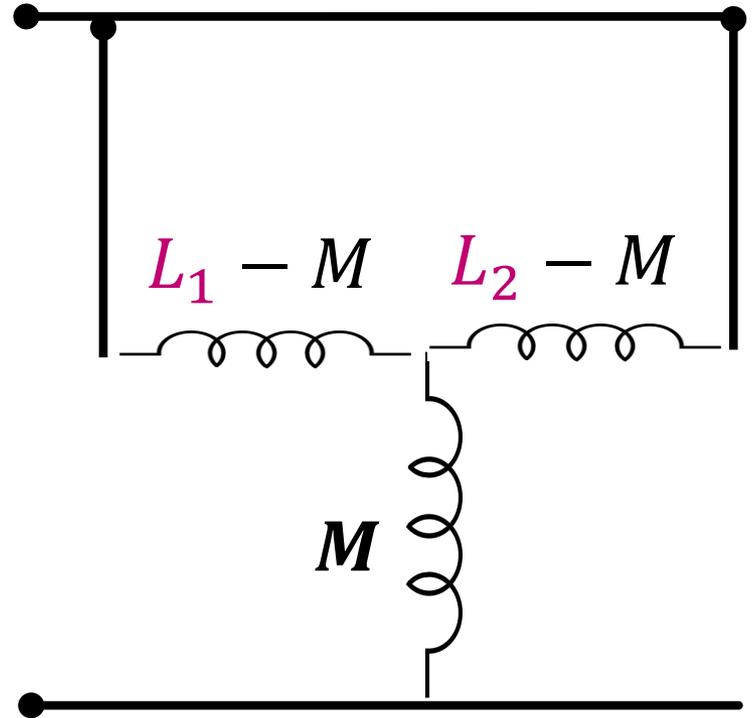
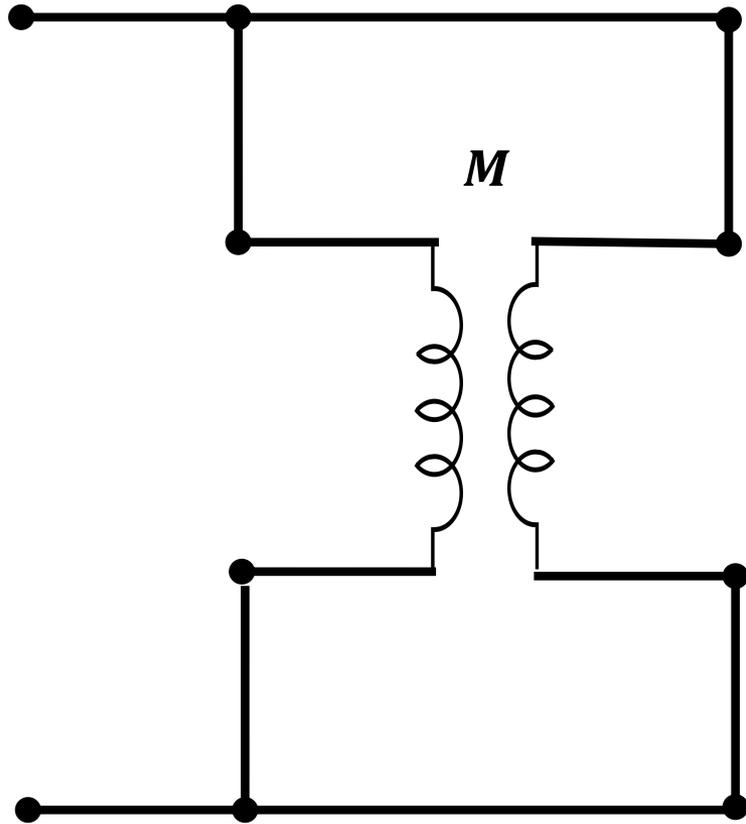
等価回路概念の導入



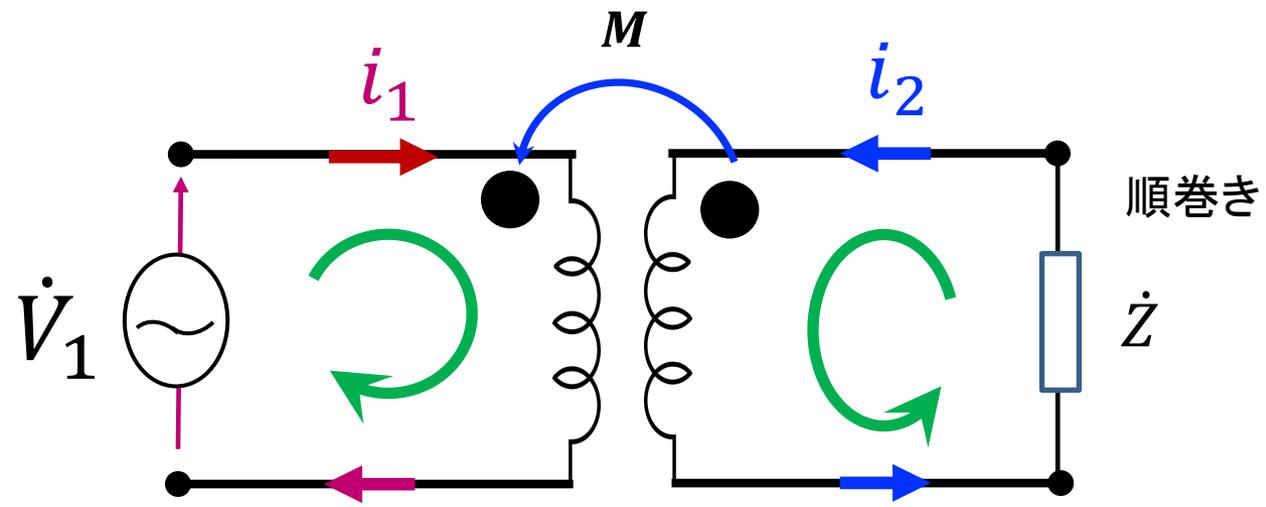
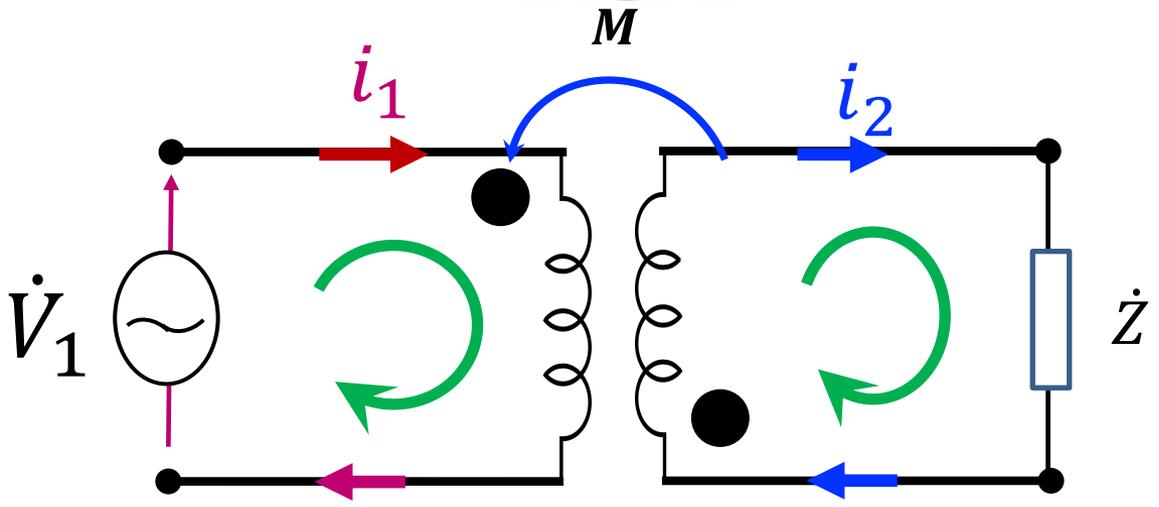
等価回路概念の導入



等価回路の例:

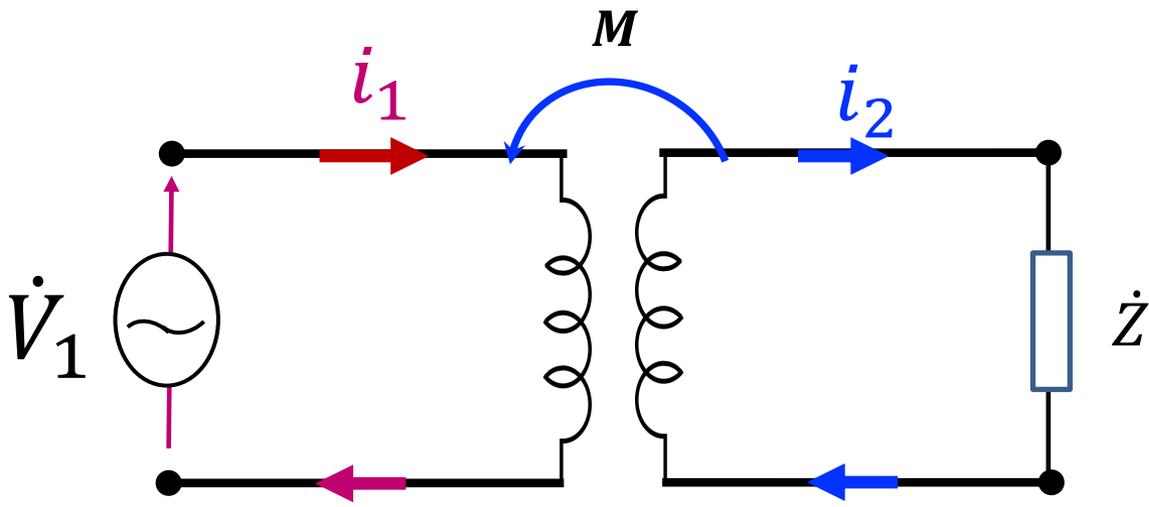


電流の方向・順巻き・逆巻き:ドット表記



変圧器と理想変圧器

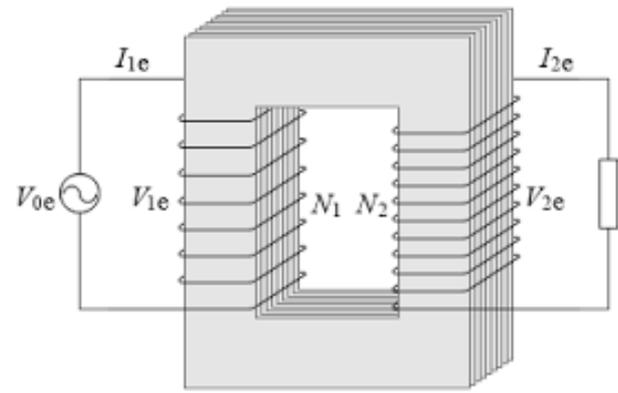
相互インダクタンスM:結合係数



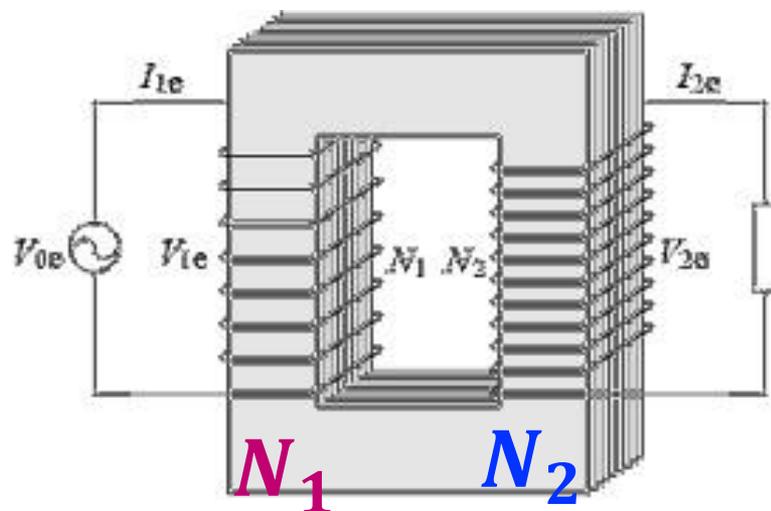
$$M = k\sqrt{L_1L_2}$$

$0 < k \leq 1$ 結合係数

$k = 1$: 変圧器結合



變壓器結合



$$\frac{N_1}{N_2} = n$$

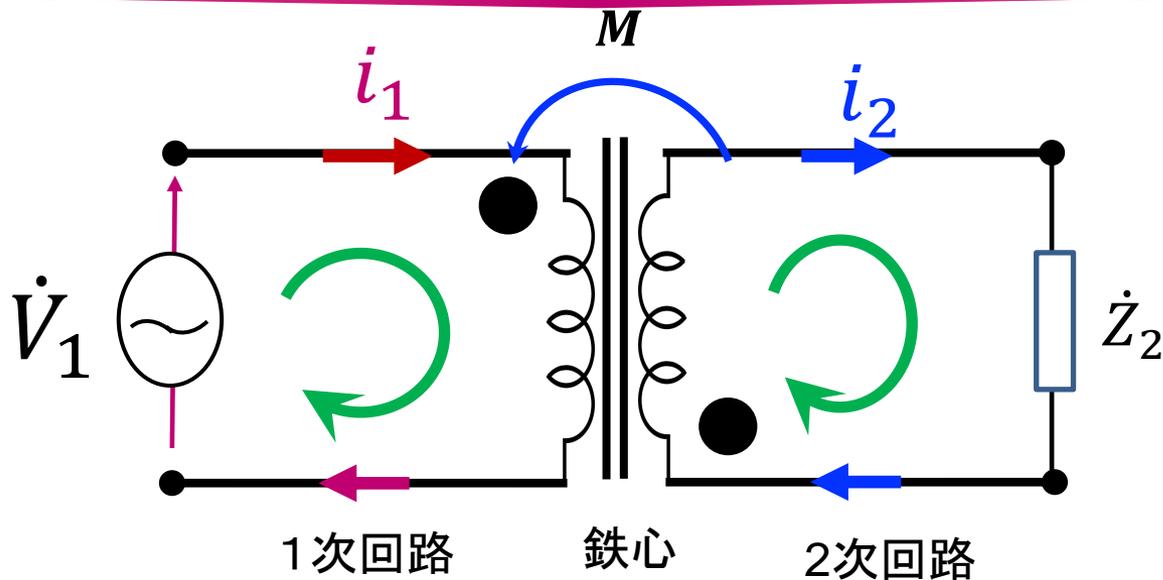
$$M = \sqrt{L_1 L_2}$$

$$L_1 = A * N_1^2$$

$$L_2 = A * N_2^2$$

$$M = A N_1 N_2 \quad \frac{L_1}{L_2} = \frac{N_1^2}{N_2^2} = n^2$$

変圧器結合回路：インピーダンス



$$\dot{Z}_1 = j\omega L_1 + \frac{\omega^2 M^2}{j\omega L_2 + \dot{Z}_2}$$

$$\dot{Z}_1 = \frac{-\omega^2 L_1 L_2 + j\omega L_1 \dot{Z}_2 + \omega^2 M^2}{j\omega L_2 + \dot{Z}_2}$$

変圧器結合回路：インピーダンス

$M = \sqrt{L_1 L_2}$ を代入する

$$\dot{Z}_1 = \frac{-\omega^2 L_1 L_2 + j\omega L_1 \dot{Z}_2 + \omega^2 L_1 L_2}{j\omega L_2 + \dot{Z}_2}$$

$$\dot{Z}_1 = \frac{j\omega L_1 \dot{Z}_2}{j\omega L_2 + \dot{Z}_2} \quad j\omega L_1 \dot{Z}_2$$

$$\dot{Y}_1 = \frac{1}{\dot{Z}_1} = \frac{j\omega L_2 + \dot{Z}_2}{j\omega L_1 \dot{Z}_2}$$

変圧器結合回路：インピーダンスとアドミタンス

$$\dot{Y}_1 = \frac{1}{\left(\frac{L_1}{L_2}\right)\dot{Z}_2} + \frac{1}{j\omega L_1}$$

$$\dot{Y}_1 = \frac{1}{n^2\dot{Z}_2} + \frac{1}{j\omega L_1}$$

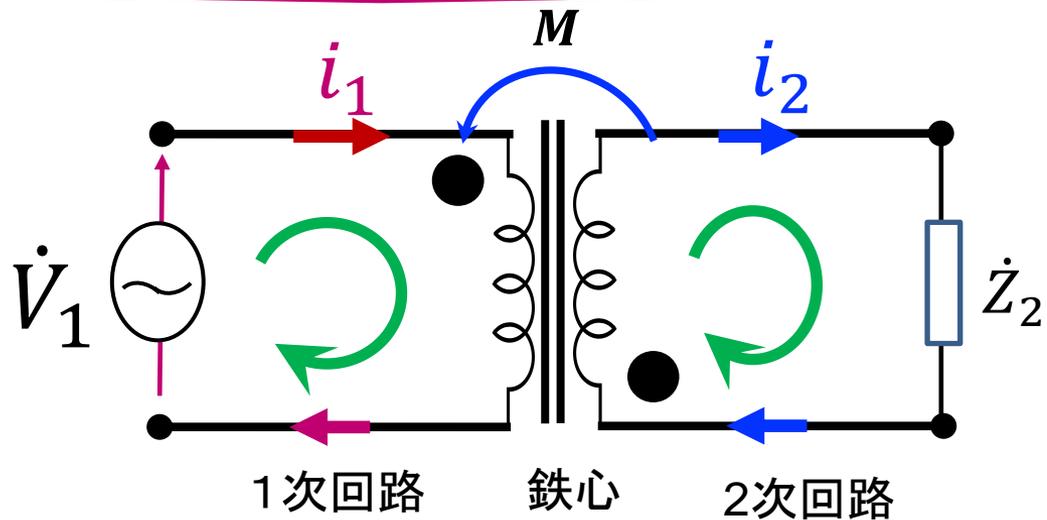
$$\dot{Z}_1 = \left(\frac{1}{n^2\dot{Z}_2} + \frac{1}{j\omega L_1}\right)^{-1}$$

$$\dot{Z}_1 = j\omega L_1 + \frac{\omega^2 M^2}{j\omega L_2 + \dot{Z}}$$

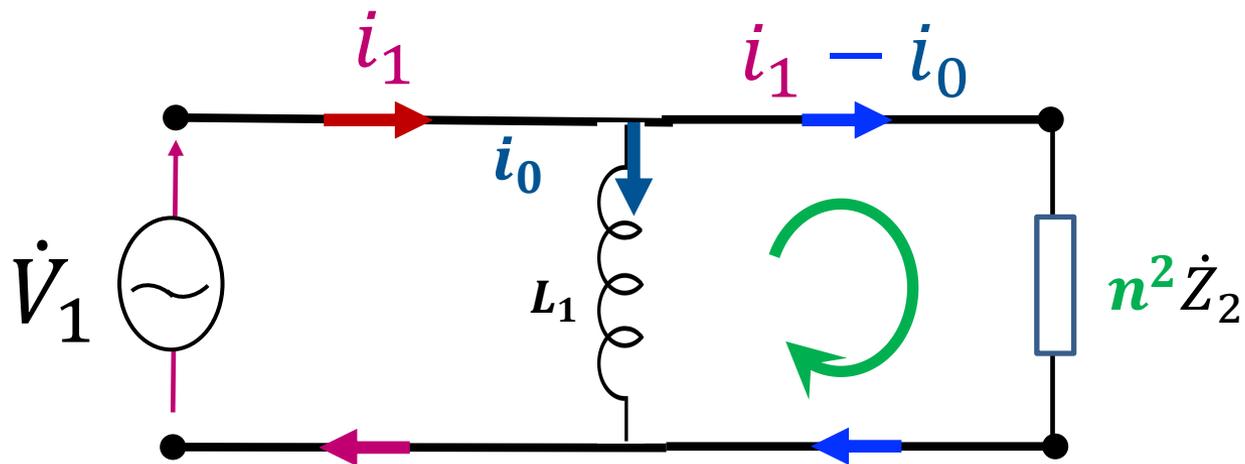
$$M = \sqrt{L_1 L_2} \qquad \frac{L_1}{L_2} = \frac{N_1^2}{N_2^2} = n^2$$

$$\dot{Z}_1 = \left(\frac{1}{n^2 \dot{Z}_2} + \frac{1}{j\omega L_1} \right)^{-1}$$

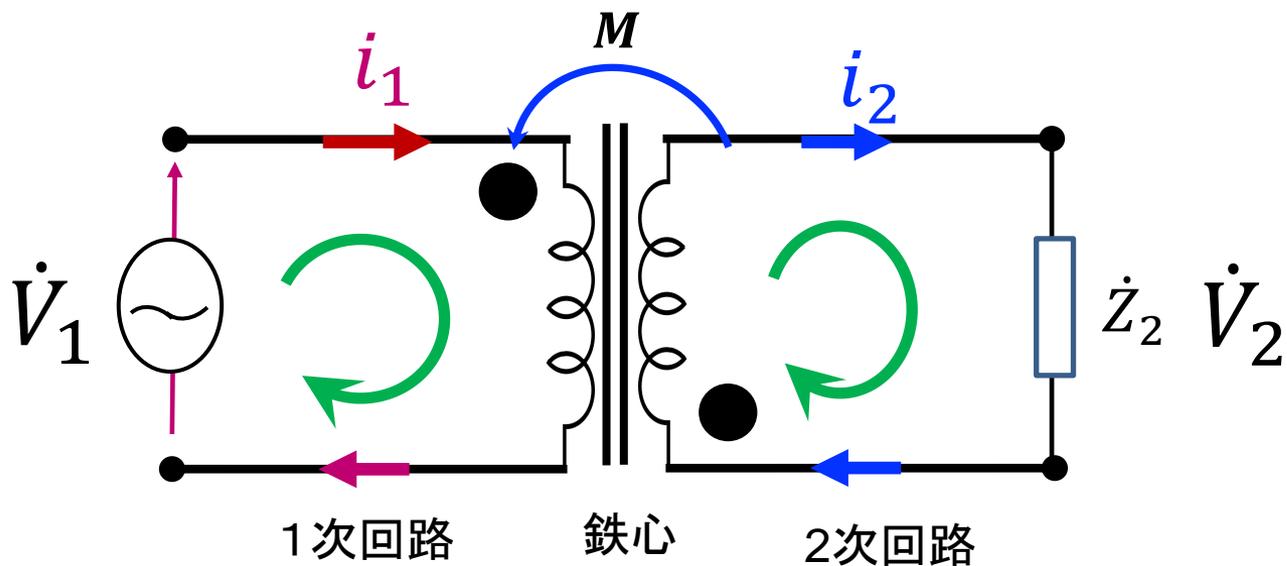
變壓器結合回路：等価回路



等
価
回
路



變壓器結合回路：電壓關係



$$i_2 = \frac{j\omega M}{j\omega L_2 + \dot{Z}_2} i_1$$

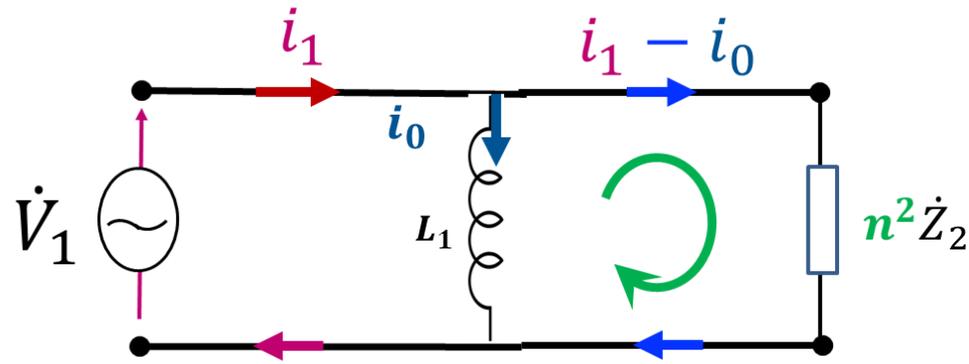
$$\dot{V}_2 = \dot{Z}_2 i_2$$

$$i_2 = \frac{j\omega M}{j\omega L_2 + \dot{Z}_2} i_1$$

$$\dot{Y}_1 = \frac{1}{n^2 \dot{Z}_2} + \frac{1}{j\omega L_1}$$

$$i_1 = \frac{j\omega L_2 + \dot{Z}_2}{j\omega L_1 \dot{Z}_2} \dot{V}_1$$

$$i_2 = \frac{j\omega M}{j\omega L_2 + \dot{Z}_2} * \frac{j\omega L_2 + \dot{Z}_2}{j\omega L_1 \dot{Z}_2} \dot{V}_1$$



$$i_2 = \frac{j\omega M}{j\omega L_1 \dot{Z}_2} \dot{V}_1$$

$$\dot{V}_2 = \dot{Z}_2 i_2$$

$$\dot{V}_2 = \dot{Z}_2 \frac{j\omega M \dot{V}_1}{\omega L_1 \dot{Z}_2}$$

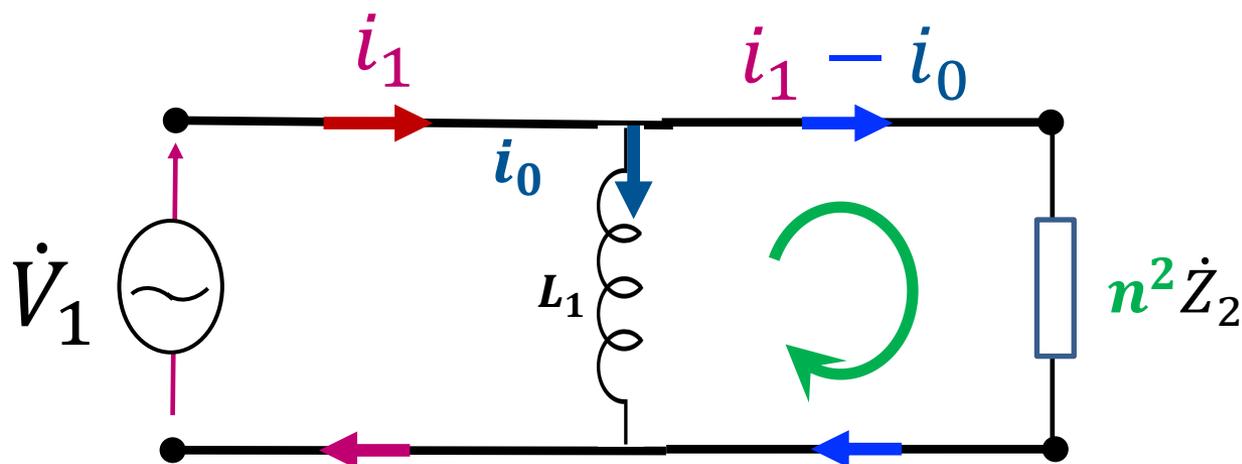
變壓器結合回路：電壓關係

$$\dot{V}_2 = \cancel{\dot{Z}_2} \frac{j\omega\sqrt{L_1L_2} \dot{V}_1}{j\omega L_1 \cancel{\dot{Z}_2}}$$

$$\dot{V}_2 = \sqrt{\frac{L_2}{L_1}} \dot{V}_1 = \frac{N_2}{N_1} \dot{V}_1 = \frac{1}{n} \dot{V}_1$$

暗記： $\dot{V}_1 = n \dot{V}_2$

変圧器結合回路：等価回路



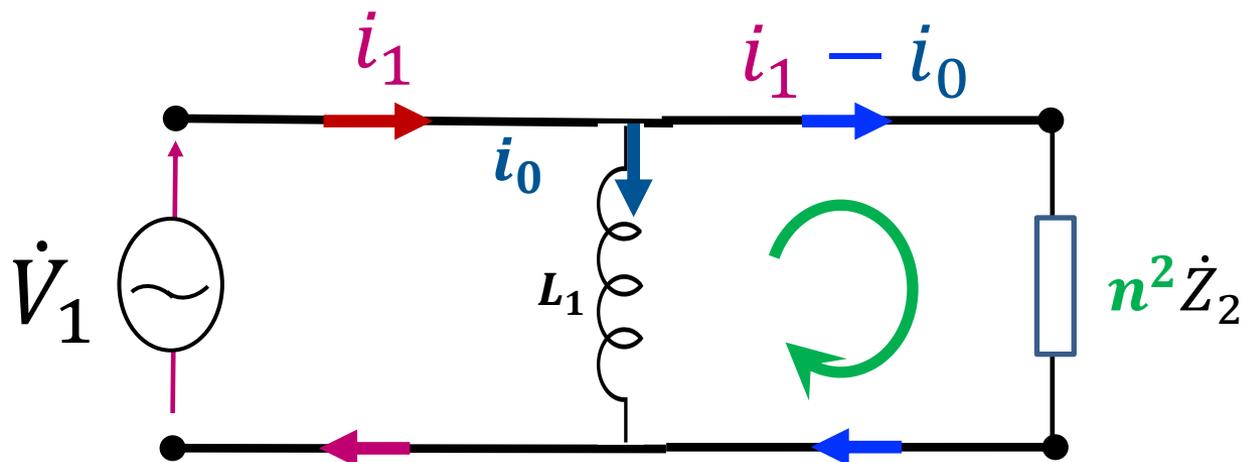
$$\frac{1}{\dot{Z}_1} = \frac{1}{n^2 \dot{Z}_2} + \frac{1}{j\omega L_1}$$

通常、変圧器は自己インダクタンス L_1 を十分大きくして、 $\omega L_1 \gg n^2 \dot{Z}_2$

$$\dot{Z}_1 = n^2 \dot{Z}_2$$

理想変圧器

変圧器結合回路：等価回路



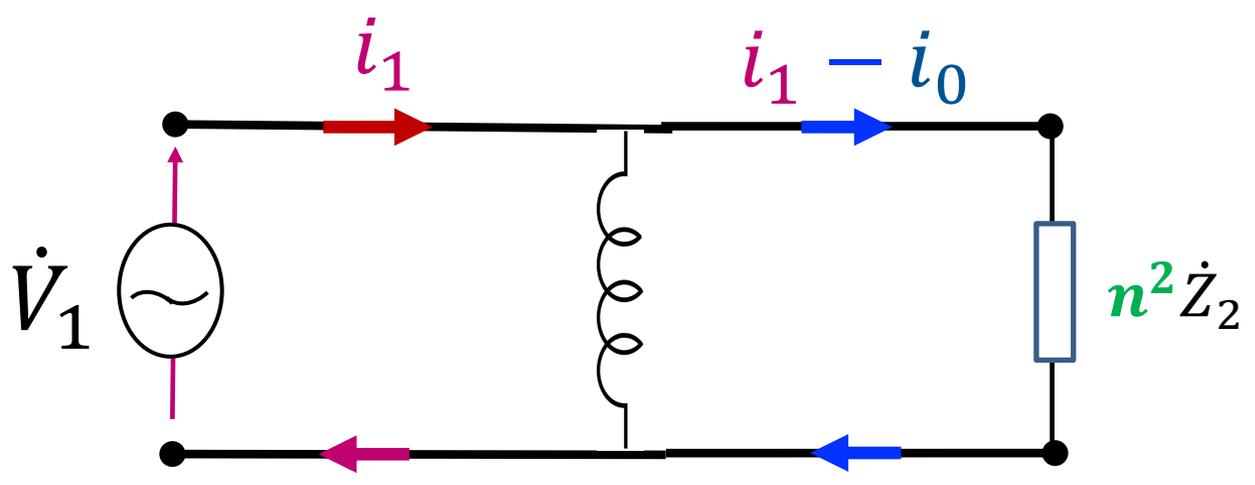
$$\frac{1}{\dot{Z}_1} = \frac{1}{n^2 \dot{Z}_2} + \frac{1}{j\omega L_1}$$

通常、変圧器は自己インダクタンス L_1 を十分大きくして、 $\omega L_1 \gg n^2 \dot{Z}_2$

$$\dot{Z}_1 = n^2 \dot{Z}_2$$

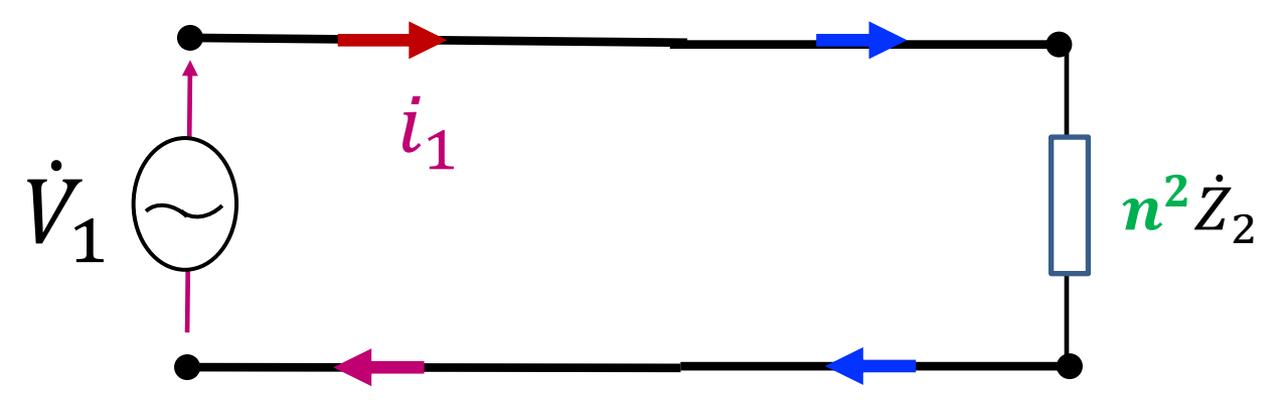
理想変圧器

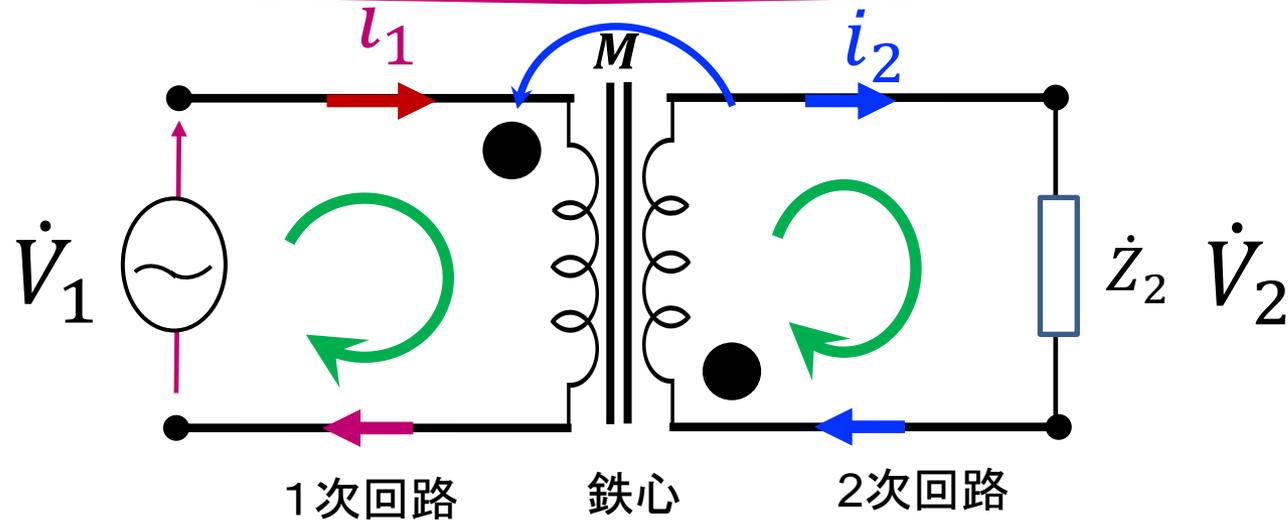
理想变压器：



理想回路

$$M = \sqrt{L_1 L_2} \quad \dot{Z}_1 = n^2 \dot{Z}_2$$

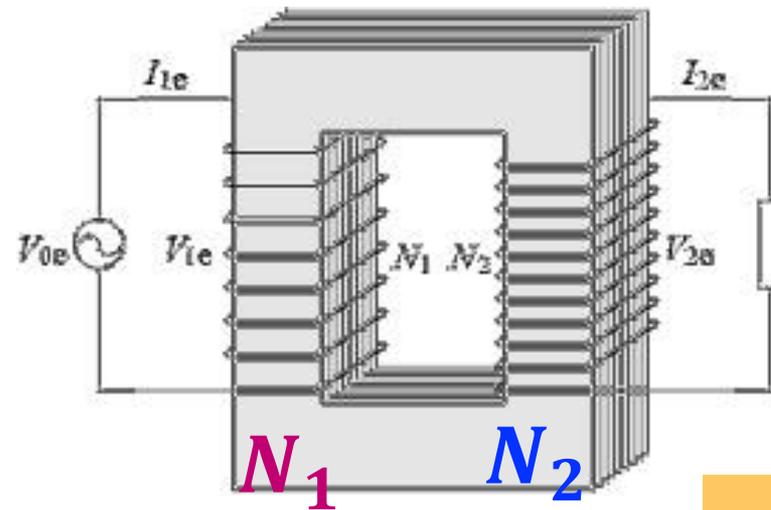




$$\dot{Z}_1 = n^2 \dot{Z}_2$$

$$i_2 = \frac{\dot{V}_2}{\dot{Z}_2} = \frac{\frac{1}{n} \dot{V}_1}{\dot{Z}_2} = \frac{\frac{1}{n} \dot{V}_1}{\frac{1}{n^2} \dot{Z}_1} = n i_1$$

理想変圧器



$$\frac{N_1}{N_2} = n$$

暗記:

$$i_2 = n i_1$$

$$\dot{V}_1 = n \dot{V}_2$$